Introduction to Nonlinear Control

Stability, control design, and estimation

Christopher M. Kellett & Philipp Braun





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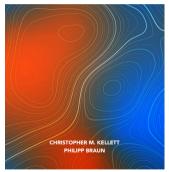
Introduction to Nonlinear Control: Stability, control design, and estimation

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Introduction to Nonlinear Control

STABILITY, CONTROL DESIGN, AND ESTIMATION



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Notation and assumptions:

- Transfer function $G : \mathbb{C} \to \mathbb{C}$.
- *G* is a rational function, i.e., there exist polynomial functions $P, Q \in \mathbb{R}[s]$ (with coefficients in \mathbb{R}) such that

$$G(s) = \frac{P(s)}{Q(s)}.$$

- *P*, *Q* are of minimal degree (i.e., they don't have common zeros)
- We assume x(0) = 0

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- Consider $\psi : \mathbb{R}_{\geq 0} \to \mathbb{R}^m$ with $\int_0^\infty \psi(t) e^{-st} dt < \infty$ for $s \in \mathbb{C}$.

Definition (Laplace transform)

Consider $\psi: \mathbb{R}_{\geq 0} \to \mathbb{R}^m$. For $s \in C \subset \mathbb{C}$ for which the integral is well-defined, the Laplace transform $\hat{\psi}: C \to \mathbb{C}^m$ of ψ is defined as

$$\hat{\psi}(s) \doteq (\mathscr{L}\psi)(s) \doteq \int_0^\infty \psi(t) e^{-st} dt.$$

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Definition (Inverse Laplace transform)

Consider $\hat{\varphi}: \mathcal{C} \to \mathbb{C}^m$ and let $\alpha \in \mathbb{R}$ such that $\alpha + j\beta \in \mathcal{C} \subset \mathbb{C}$ for all $\beta \in \mathbb{R}$. Then the inverse Laplace transform $\varphi: \mathbb{R}_{\geq 0} \to \mathbb{R}^m$ of $\hat{\varphi}$ is defined as

$$\begin{aligned} \varphi(t) &= (\mathscr{L}^{-1}\hat{\varphi})(t) = \frac{1}{2\pi j} \int_{\alpha-j\infty}^{\alpha+j\infty} e^{st} \hat{\varphi}(s) \ ds \\ &= \frac{e^{\alpha t}}{2\pi j} \int_{-\infty}^{\infty} e^{jwt} \hat{\varphi}(\alpha+jw) \ dw. \end{aligned}$$

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Notation and assumptions:

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- We assume x(0) = 0
- Consider $\psi : \mathbb{R}_{\geq 0} \to \mathbb{R}^m$ with $\int_0^\infty \psi(t) e^{-st} dt < \infty$ for $s \in \mathbb{C}$.

Proposition (Laplace transform properties)

Consider the signals

$$\varphi, \varphi_1, \varphi_2 : \mathbb{R}_{\geq 0} \to \mathbb{R}^m$$

in the time domain and constants

 $a \in \mathbb{R}_{>0}, a_1, a_2 \in \mathbb{R}.$

Then the Laplace transform satisfies

$$\begin{split} \mathscr{L}^{-1}\mathscr{L}\varphi(t) &= \varphi(t),\\ \mathscr{L}(a_1\varphi_1 + a_2\varphi_2)(s) &= a_1\hat{\varphi}_1(s) + a_2\hat{\varphi}_2(s),\\ \mathscr{L}(\varphi(a \cdot))(s) &= \frac{1}{a}\hat{\varphi}\left(\frac{s}{a}\right),\\ \mathscr{L}(\varphi(\cdot - a))(s) &= e^{-sa}\hat{\varphi}\left(s\right),\\ \mathscr{L}\left(\frac{d^k}{dt^k}\varphi\right)(s) &= s^k\hat{\varphi}(s) - \sum_{j=1}^{k-1}s^{j-1}\frac{d^{k-1-j}}{dt^{k-1-j}}\varphi(0),\\ \mathscr{L}\left(\int_0^{\cdot}\varphi(\tau) d\tau\right)(s) &= \frac{1}{s}\hat{\varphi}(s). \end{split}$$

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Application of the Laplace transform:

 $s\hat{x}(s) - x(0) = A\hat{x}(s) + b\hat{u}(s), \qquad \hat{y}(s) = c\hat{x}(s) + d\hat{u}(s)$

Rearrange the terms (x(0) = 0):

 $\hat{y}(s) = (c(sI - A)^{-1}b + d) \hat{u}(s)$

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Identify input output relationship:

$$G(s) = \frac{\hat{y}(s)}{\hat{u}(s)} = c(sI - A)^{-1}b + d \tag{1}$$

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Identify input output relationship:

$$G(s) = \frac{\hat{y}(s)}{\hat{u}(s)} = c(sI - A)^{-1}b + d$$
(1)

Definition (Realization)

Consider a transfer function G(s) and assume that (1) is satisfied for (A, b, c, d). Then G(s) is called realizable and the quadruple (A, b, c, d) is called a realization of G(s).

Proposition (Laplace transform properties)

Consider the signals $\varphi, \varphi_1, \varphi_2 : \mathbb{R}_{\geq 0} \to \mathbb{R}^m$ in the time domain and constants $a \in \mathbb{R}_{>0}, a_1, a_2 \in \mathbb{R}$. Then the Laplace transform and its inverse satisfy the following properties:

$$\begin{split} \mathscr{L}^{-1}\mathscr{L}\varphi(t) &= \varphi(t), \\ \mathscr{U}(a_1\varphi + a_2\varphi_2)(s) &= a_1\hat{\varphi}_1(s) + a_2\hat{\varphi}_2(s), \\ \mathscr{L}(\varphi(a \cdot))(s) &= \frac{1}{a}\hat{\varphi}\left(\frac{s}{a}\right), \\ \mathscr{L}(\varphi(\cdot - a))(s) &= e^{-sa}\hat{\varphi}\left(s\right), \\ \mathscr{L}\left(\frac{d^k}{dt^k}\varphi\right)(s) &= s^k\hat{\varphi}(s) - \sum_{j=1}^{k-1} s^{j-1} \frac{d^{k-1-j}}{dt^{k-1-j}}\varphi(0), \\ \mathscr{U}\left(\int_0^{\cdot}\varphi(\tau) d\tau\right)(s) &= \frac{1}{s}\hat{\varphi}(s). \end{split}$$

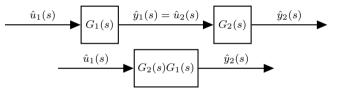
Consider two systems:

 $\hat{y}_1(s) = G(s)\hat{u}_1(s)$ $\hat{y}_2(s) = G(s)\hat{u}_2(s)$

Cascade interconnection

$$\hat{y}_2(s) = G_2(s)G_1(s)\hat{u}_1(s)$$

Cascade interconnection $\hat{u}_2(s) = \hat{y}_1(s)$

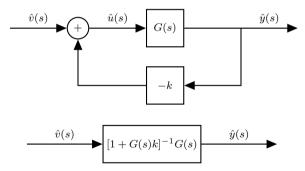


System Interconnections in the Frequency Domain

Consider:

$$\hat{y}(s) = G(s)\hat{u}(s)$$
$$\hat{u}(s) = \hat{v}(s) - k\hat{y}(s)$$

Feedback interconnection:

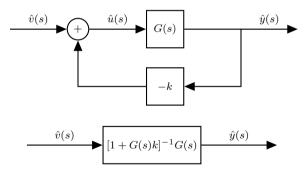


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Feedback interconnection:



Stability and robustness analysis tools:

- Bode diagram
- Nyquist diagram (and Nyquist criterion)

Part I: Dynamical Systems

- 1. Nonlinear Systems -Fundamentals & Examples
- 2. Nonlinear Systems Stability Notions
- 3. Linear Systems and Linearization
- 4. Frequency Domain Analysis
- 5. Discrete Time Systems
- 6. Absolute Stability
- 7. Input-to-State Stability

Part II: Controller Design

- 8. LMI Based Controller and Antiwindup Designs
- 9. Control Lyapunov Functions
- 10. Sliding Mode Control
- 11. Adaptive Control
- 12. Introduction to Differential Geometric Methods
- 13. Output Regulation
- 14. Optimal Control
- 15. Model Predictive Control

Part III: Observer Design & Estimation

- Observer Design for Linear Systems
- 17. Extended & Unscented Kalman Filter & Moving Horizon Estimation
- Observer Design for Nonlinear Systems